UNIVERSITY OF SWAZILAND FACULTY OF SCIENCE DEPARTMENT OF ELECTRONIC ENGINEERING

MAIN EXAMINATION

December

2006

TITLE OF PAPER: ADVANCED CONTROL SYSTEMS

COURSE NUMBER: EIN530

TIME ALLOWED: THREE HOURS

INSTRUCTIONS: ANSWER QUESTION 1 AND ANY OTHER THREE QUESTIONS

EACH QUESTION CARRIES 25 MARKS

MARKS FOR DIFFERENT SECTIONS ARE SHOWN IN THE RIGHT-HAND MARGIN

THIS PAPER HAS 7 PAGES, INCLUDING THIS PAGE

DO NOT OPEN THE PAPER UNTIL PERMISSION HAS BEEN GIVEN BY THE INVIGILATOR

Partial Table of z- and s-Transforms

	· f(t) /	F(s) F(z)	f(kt)
	u(t)	$\frac{1}{s}$ $\frac{z}{z-1}$	u(kT)
2.	+ t	$\frac{1}{s^2} \qquad \frac{Tz}{(z-1)^2}$	kT
3.	t^n	$\frac{n!}{s^{n+1}} \qquad \lim_{a \to 0} (-1)^n \frac{d^n}{da^n} \left[\frac{z}{z - e^{-aT}} \right]$	$(kT)^n$
4.	e^{-at}	$\frac{1}{s+a} \qquad \qquad \frac{z}{z-e^{-aT}}$	e^{-akT}
5.	$t^n e^{-at}$	$\frac{n!}{(s+a)^{n+1}} \qquad (-1)^n \frac{d^n}{da^n} \left[\frac{z}{z-e^{-aT}} \right]$	$(kT)^n e^{-akT}$
6.	$\sin \omega t$	$\frac{\omega}{s^2 + \omega^2} \qquad \frac{z \sin \omega T}{z^2 - 2z \cos \omega T + 1}$	sin ω kT
7.	$\cos \omega t$	$\frac{s}{s^2 + \omega^2} \qquad \frac{z(z - \cos \omega T)}{z^2 - 2z \cos \omega T + 1}$	cos ω kT
8.	$e^{-at}\sin\omega t$	$\frac{\omega}{(s+a)^2 + \omega^2} = \frac{ze^{-aT}\sin\omega T}{z^2 - 2ze^{-aT}\cos\omega T + e^{-2aT}}$	$e^{-ikT}\sin\omega kT$
9.	$e^{-at}\cos\omega t$	$\frac{s+a}{(s+a)^2+\omega^2} \qquad \frac{z^2-ze^{-aT}\cos\omega T}{z^2-2ze^{-aT}\cos\omega T+e^{-2aT}}$	$e^{-akT}\cos\omega kT$
0.		<u>Z</u> Z + a	a ^K Cos Kπ

z-Transform Theorems

	Theorem	Name
1.	$z\{af(t)\} = aF(z)$	Linearity theorem
2.	$z\{f_1(t) + f_2(t)\} = F_1(z) + F_2(z)$	Linearity theorem
.3.	$z\{e^{-at}f(t)\} = F(e^{aT}z)$	Complex differentiation
4.	$z\{f(t-nT)\} = z^{-n}F(z)$	Real translation
5.	$z\{tf(t)\} = -Tz\frac{dF(z)}{dz}$	Complex differentiation
6.	$f(0) \doteq \lim_{z \to \infty} F(z)$	Initial value theorem
7.	$f(\infty) = \lim_{z \to 1} (1 - z^{-1}) F(z)$	Final value theorem

Note: kT may be substituted for t in the table.

State the definitions and give one major difficulty of each of the following advanced control methods :

Adaptive control, Robust control, Predictive control, Optimal control, and Fuzzy control

A) A system is described by the matrix equations

$$\dot{x} = \begin{bmatrix} 0 & 1 \\ -1 & -2 \end{bmatrix} x + \begin{bmatrix} 1 \\ -1 \end{bmatrix} u$$

and $y(t) = x_1(t)$. Determine whether the system is controllable and observable.

[8 marks]

B) For a system described by the matrix equations

$$\dot{x} = \begin{bmatrix} 0 & 1 \\ -2 & -3 \end{bmatrix} x + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u$$

$$u = -(k_1 x_1 + k_2 x_2)$$

Determine the values of k1 and k2 so that a full-state variable feedback design is achieved to meet a settling time requirement of 0.8 seconds with $\zeta = 0.884$.

[17 marks]

Consider a system with

$$\dot{x} = \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix} x + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u$$

$$u = -(x_1 + k_1 x_2)$$

 $u = -(x_1 + k_1 x_2)$ and a performance index of $J = \int\limits_0^\infty x^T x dt$.

- A) Determine the optimum value of k_l and the minimum value of J when $x^T = \begin{bmatrix} 1 \end{bmatrix}$
- [19 marks] B) Plot the performance index versus the gain k_1 [6 marks]

A unity feedback control system has a loop transfer function

$$G(s) = \frac{10^4}{(s+10^3)}$$

It is desired that the phase margin of this system be at least 40°. Design a lead compensator using Bode diagrams to achieve this specification. [25 marks]

Figure 5 shows a satellite tracking system. It is desired to design the controller D(z), so that this system is deadbeat and the steady state error to a step input R(z) is zero. The sampling interval is T = 0.01.

[25 marks]

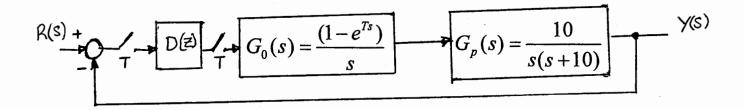


Figure 5