

**UNIVERSITY OF SWAZILAND**  
**FACULTY OF SCIENCE & ENGINEERING**  
**DEPARTMENT OF ELECTRICAL & ELECTRONIC ENGINEERING**

**SIGNALS AND SYSTEMS II**

**COURSE CODE - EE332**

**MAIN EXAMINATION**

**MAY 2016**

**DURATION OF THE EXAMINATION - 3 HOURS**

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**INSTRUCTIONS TO CANDIDATES**

1. There are FIVE questions in this paper. Answer any FOUR questions.
3. Show all your steps clearly in any calculations/work.
4. State clearly any assumptions made.
5. Start each new question on a fresh page.
6. Useful Fourier Transform and Z-Transform properties are attached.
7. Make sure that this exam contains 8 pages including this one.

**DO NOT OPEN THIS PAPER UNTIL PERMISSION HAS BEEN GIVEN BY THE INVIGILATOR.**

**QUESTION ONE (25 marks)**

Let  $x(t)$  be a possibly complex valued time signal with Fourier transform  $X(\omega)$ . Express the Fourier transforms of the following time signals in terms of  $X(\omega)$ .

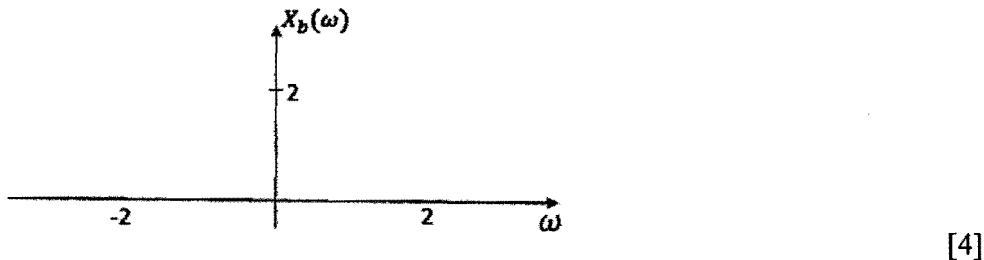
- a.  $x^*(t)$  [3]
- b.  $\text{Im}\{x(t)\}$  [6]
- c.  $x(t - 2) + x^*(-t + 2)$  [8]
- d.  $e^{j4\pi t}x(3t - 3)$  [8]

**QUESTION TWO (25 marks)**

Find the signal corresponding to the following Fourier transforms.

a.  $X_a(\omega) = \frac{1}{2+j\omega}$  [3]

b.



c.  $X_c(e^{j\omega}) = \frac{e^{-j2(\omega-\frac{\pi}{4})}}{2-e^{-j(\omega-\frac{\pi}{4})}}$  [8]

d.  $X_d(\omega) = X_a(\omega)X_b(\omega)$  [10]

**QUESTION THREE (25 marks)**

A causal and stable LTI system has a property that

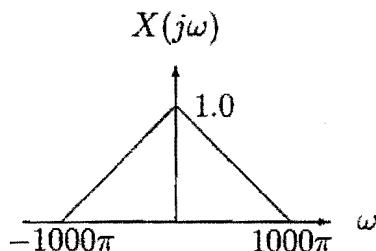
$$\left(\frac{4}{5}\right)^n u(n) \rightarrow n \left(\frac{4}{5}\right)^n u(n).$$

- a. Define the discrete time Fourier transform (DTFT) pairs and indicate how DTFT is related to continuous time Fourier transform (CTFT). [5]
- b. Determine the frequency response  $H(e^{j\omega})$  for the system. [10]

- c. Determine a difference equation relating any input  $x(n)$  and the corresponding output  $y(n)$ . [10]

**QUESTION FOUR (25 marks)**

1. State sampling theorem. Define Nyquist rate and Aliasing? [5]
2. Let  $x(t)$  be a real-valued signal for which  $X(\omega) = 0$  when  $|\omega| > 1000\pi$ . Amplitude modulation is performed to produce the signal  $g(t) = x(t) \cos(1000\pi t)$ . Assume the spectrum of  $x(t)$  is real valued and as given in the figure.
  - a. Find and plot the real and imaginary parts of  $G(\omega)$ , the spectrum of  $g(t)$ . [8]
  - b. Find and plot the real and imaginary parts of  $W(\omega)$ , the spectrum of  $w(t)$ . [12]



**QUESTION FIVE (25 marks)**

Consider the two discrete-time sequences.

$$x_1(n) = \left(\frac{1}{2}\right)^n u(n-1) \text{ and } x_2(n) = \left(\frac{1}{3}\right)^n u(n+1)$$

where  $u(n)$  is the unit step function. A third signal is obtained using the convolution sum, exam.  $x(n) = x_1(n) * x_2(n) * x_2(n)$ .

- a. Compute the z-transform of  $x_1(n)$ . [5]
- b. Compute the z-transform of  $x_2(n)$ . [5]
- c. Compute the z-transform of  $x(n)$ . [5]
- d. Derive  $x(n)$  from the z-transform obtained in part (c). [10]

TABLE I Fourier Transforms

No.	$x(t)$	$X(\omega)$	
1	$e^{-at}u(t)$	$\frac{1}{a + j\omega}$	$a > 0$
2	$e^{at}u(-t)$	$\frac{1}{a - j\omega}$	$a > 0$
3	$e^{-a t }$	$\frac{2a}{a^2 + \omega^2}$	$a > 0$
4	$te^{-at}u(t)$	$\frac{1}{(a + j\omega)^2}$	$a > 0$
5	$t^n e^{-at}u(t)$	$\frac{n!}{(a + j\omega)^{n+1}}$	$a > 0$
6	$\delta(t)$	1	
7	1	$2\pi\delta(\omega)$	
8	$e^{j\omega_0 t}$	$2\pi\delta(\omega - \omega_0)$	
9	$\cos \omega_0 t$	$\pi[\delta(\omega - \omega_0) + \delta(\omega + \omega_0)]$	
10	$\sin \omega_0 t$	$j\pi[\delta(\omega + \omega_0) - \delta(\omega - \omega_0)]$	
11	$u(t)$	$\pi\delta(\omega) + \frac{1}{j\omega}$	
12	$\operatorname{sgn} t$	$\frac{2}{j\omega}$	
13	$\cos \omega_0 t u(t)$	$\frac{\pi}{2}[\delta(\omega - \omega_0) + \delta(\omega + \omega_0)] + \frac{j\omega}{\omega_0^2 - \omega^2}$	
14	$\sin \omega_0 t u(t)$	$\frac{\pi}{2j}[\delta(\omega - \omega_0) - \delta(\omega + \omega_0)] + \frac{\omega_0}{\omega_0^2 - \omega^2}$	
15	$e^{-at} \sin \omega_0 t u(t)$	$\frac{\omega_0}{(a + j\omega)^2 + \omega_0^2}$	$a > 0$
16	$e^{-at} \cos \omega_0 t u(t)$	$\frac{a + j\omega}{(a + j\omega)^2 + \omega_0^2}$	$a > 0$
17	$\operatorname{rect}\left(\frac{t}{\tau}\right)$	$\tau \operatorname{sinc}\left(\frac{\omega\tau}{2}\right)$	
18	$\frac{W}{\pi} \operatorname{sinc}(Wt)$	$\operatorname{rect}\left(\frac{\omega}{2W}\right)$	
19	$\Delta\left(\frac{t}{\tau}\right)$	$\frac{\tau}{2} \operatorname{sinc}^2\left(\frac{\omega\tau}{4}\right)$	
20	$\frac{W}{2\pi} \operatorname{sinc}^2\left(\frac{Wt}{2}\right)$	$\Delta\left(\frac{\omega}{2W}\right)$	
21	$\sum_{n=-\infty}^{\infty} \delta(t - nT)$	$\omega_0 \sum_{n=-\infty}^{\infty} \delta(\omega - n\omega_0)$	$\omega_0 = \frac{2\pi}{T}$
22	$e^{-t^2/2\sigma^2}$	$\sigma\sqrt{2\pi}e^{-\sigma^2\omega^2/2}$	

Table 2 Fourier Transform Operations

Operation	$x(t)$	$X(\omega)$
Scalar multiplication	$kx(t)$	$kX(\omega)$
Addition	$x_1(t) + x_2(t)$	$X_1(\omega) + X_2(\omega)$
Conjugation	$x^*(t)$	$X^*(-\omega)$
Duality	$X(t)$	$2\pi x(-\omega)$
Scaling ( $a$ real)	$x(at)$	$\frac{1}{ a } X\left(\frac{\omega}{a}\right)$
Time shifting	$x(t - t_0)$	$X(\omega)e^{-j\omega t_0}$
Frequency shifting ( $\omega_0$ real)	$x(t)e^{j\omega_0 t}$	$X(\omega - \omega_0)$
Time convolution	$x_1(t) * x_2(t)$	$X_1(\omega)X_2(\omega)$
Frequency convolution	$x_1(t)x_2(t)$	$\frac{1}{2\pi} X_1(\omega) * X_2(\omega)$
Time differentiation	$\frac{d^n x}{dt^n}$	$(j\omega)^n X(\omega)$
Time integration	$\int_{-\infty}^t x(u) du$	$\frac{X(\omega)}{j\omega} + \pi X(0)\delta(\omega)$

## DISCRETE-TIME FOURIER TRANSFORM

### A. Properties of the discrete-time Fourier transform

<i>Non-periodic signal</i>	<i>Fourier transform</i>
$x[n] = \frac{1}{2\pi} \int_{2\pi} X(e^{j\omega}) e^{j\omega n} d\omega$	$X(e^{j\omega}) \triangleq \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n}$
$\left. \begin{array}{l} x[n] \\ y[n] \end{array} \right\}$	$\left. \begin{array}{l} X(e^{j\omega}) \\ Y(e^{j\omega}) \end{array} \right\}$ Periodic with period $2\pi$
$ax[n] + by[n]$	$aX(e^{j\omega}) + bY(e^{j\omega})$
$x[n - n_0]$	$e^{-j\omega n_0} X(e^{j\omega})$
$e^{j\omega_0 n} x[n]$	$X(e^{j(\omega - \omega_0)})$
$x^*[n]$	$X^*(e^{j(-\omega)})$
$x[-n]$	$X(e^{j(-\omega)})$
$x_{(m)}[n] = \left\{ \begin{array}{ll} x[n/m], & n \text{ multiple of } m \\ 0, & n \text{ not multiple of } m \end{array} \right.$	$X(e^{j(m\omega)})$
$x[n] * y[n]$	$X(e^{j\omega})Y(e^{j\omega})$
$x[n]y[n]$	$\frac{1}{2\pi} \int_{2\pi} X(e^{j\theta})Y(e^{j(\omega-\theta)}) d\theta$
$x[n] - x[n - 1]$	$(1 - e^{j\omega}) X(e^{j\omega})$
$\sum_{k=-\infty}^n x[k]$	$\frac{1}{1 - e^{j\omega}} X(e^{j\omega}) + \pi X(0) \sum_{k=-\infty}^{\infty} \delta(\omega - 2\pi k)$
$nx[n]$	$j \frac{d}{d\omega} X(e^{j\omega})$
<i>If <math>x[n]</math> is real valued then</i>	
$x[n]$	$\left\{ \begin{array}{l} X(e^{j\omega}) = X^*(e^{j(-\omega)}) \\ \Re\{X(e^{j\omega})\} = \Re\{X(e^{j(-\omega)})\} \\ \Im\{X(e^{j\omega})\} = -\Im\{X(e^{j(-\omega)})\} \\  X(e^{j\omega})  =  X(e^{j(-\omega)})  \\ \arg\{X(e^{j\omega})\} = -\arg\{X(e^{j(-\omega)})\} \end{array} \right.$
$x_e[n] = \mathcal{E}\{x[n]\}$	$\Re\{X(e^{j\omega})\}$
$x_o[n] = \mathcal{O}\{x[n]\}$	$j\Im\{X(e^{j\omega})\}$
<i>Parsevals relation for non-periodic signals</i>	
$\sum_{n=-\infty}^{\infty}  x[n] ^2 = \frac{1}{2\pi} \int_{2\pi}  X(e^{j\omega}) ^2 d\omega$	

B. Discrete-time Fourier transform table

$x[n]$	$X(e^{j\omega})$
$\delta[n]$	1
$\delta[n - n_0]$	$e^{-j\omega n_0}$
$\sum_{k=-\infty}^{\infty} \delta(n - kN)$	$\frac{2\pi}{N} \sum_{k=-\infty}^{\infty} \delta\left(\omega - \frac{2\pi k}{N}\right)$
1	$2\pi \sum_{k=-\infty}^{\infty} \delta(\omega - 2\pi k)$
$e^{j\omega_0 n}$	$2\pi \sum_{k=-\infty}^{\infty} \delta(\omega - \omega_0 - 2\pi k)$
$\cos \omega_0 n$	$\pi \sum_{k=-\infty}^{\infty} [\delta(\omega - \omega_0 - 2\pi k) + \delta(\omega + \omega_0 - 2\pi k)]$
$\sin \omega_0 n$	$\frac{\pi}{j} \sum_{k=-\infty}^{\infty} [\delta(\omega - \omega_0 - 2\pi k) - \delta(\omega + \omega_0 - 2\pi k)]$
$u[n]$	$\frac{1}{1 - e^{-j\omega}} + \pi \sum_{k=-\infty}^{\infty} \delta(\omega - 2\pi k)$
$a^n u(n), \quad  a  < 1$	$\frac{1}{1 - ae^{-j\omega}}$
$(n+1)a^n u[n], \quad  a  < 1$	$\frac{1}{(1 - ae^{-j\omega})^2}$
$\frac{(n+m-1)!}{n!(m-1)!} a^n u[n], \quad  a  < 1$	$\frac{1}{(1 - ae^{-j\omega})^m}$
$\frac{1}{1 - a^2} a^{ n }, \quad  a  < 1$	$\frac{1}{1 + a^2 - 2a \cos \omega}$
$\begin{cases} 1, &  n  \leq N_1 \\ 0, & N_1 <  n  \leq \frac{N}{2} \end{cases}$ period $N$	$2\pi \sum_{k=-\infty}^{\infty} a_k \delta\left(\omega \frac{2\pi k}{N}\right)$
$\begin{cases} 1, &  n  \leq N_1 \\ 0, &  n  > N_1 \end{cases}$	$\frac{\sin \omega (N_1 + \frac{1}{2})}{\sin \frac{\omega}{2}}$
$\begin{cases} \frac{\sin Wn}{W} = \frac{W}{\pi} \text{sinc} \frac{Wn}{\pi} \\ 0 < \frac{\pi n}{W} < \pi \end{cases}$	$\begin{cases} 1, &  \omega  \leq W \\ 0, & W <  \omega  \leq \pi \end{cases}$ period $2\pi$

### Table of Z-Transforms

Line No.	$x(n), n \geq 0$	z-Transform $X(z)$	Region of Convergence
1	$x(n)$	$\sum_{n=0}^{\infty} x(n)z^{-n}$	
2	$\delta(n)$	1	$ z  > 0$
3	$a^n u(n)$	$\frac{az}{z - 1}$	$ z  > 1$
4	$nu(n)$	$\frac{z}{(z - 1)^2}$	$ z  > 1$
5	$n^2 u(n)$	$\frac{z(z + 1)}{(z - 1)^3}$	$ z  > 1$
6	$a^n u(n)$	$\frac{z}{z - a}$	$ z  >  a $
7	$e^{-an} u(n)$	$\frac{z}{(z - e^{-a})}$	$ z  > e^{-a}$
8	$na^n u(n)$	$\frac{az}{(z - a)^2}$	$ z  >  a $
9	$\sin(\omega n)u(n)$	$\frac{z \sin(\omega)}{z^2 - 2z \cos(\omega) + 1}$	$ z  > 1$
10	$\cos(\omega n)u(n)$	$\frac{z[z - \cos(\omega)]}{z^2 - 2z \cos(\omega) + 1}$	$ z  > 1$
11	$a^n \sin(bn)u(n)$	$\frac{[a \sin(b)]z}{z^2 - [2z \cos(b)]z + a^2}$	$ z  >  a $
12	$a^n \cos(bn)u(n)$	$\frac{z[z - a \cos(b)]}{z^2 - [2z \cos(b)]z + a^2}$	$ z  >  a $
13	$e^{-an} \sin(bn)u(n)$	$\frac{[e^{-a} \sin(b)]z}{z^2 - [2e^{-a} \cos(b)]z + e^{-2a}}$	$ z  > e^{-a}$
14	$e^{-an} \cos(bn)u(n)$	$\frac{z[z - e^{-a} \cos(b)]}{z^2 - [2e^{-a} \cos(b)]z + e^{-2a}}$	$ z  > e^{-a}$

### Properties of Z-Transforms

Linearity:  $ax_1[k] + bx_2[k] \Leftrightarrow aX_1(z) + bX_2(z)$

Time Reversal:  $x[-k] \Leftrightarrow X(1/z)$

Summation:  $\sum_{n=-\infty}^k x[n] \Leftrightarrow \frac{zX(z)}{z - 1}$

Initial Value:  $x[0] = \lim_{z \rightarrow \infty} X(z)$

Final Value:  $x[\infty] = \lim_{z \rightarrow 1} (z - 1)X(z)$

Convolution:  $x[k] * h[k] \Leftrightarrow X(z)H(z)$

Differencing:  $x[k] - x[k - 1] \Leftrightarrow (1 - z^{-1})X(z)$

Differentiation:  $-kx[k] \Leftrightarrow z \frac{d}{dz} X(z)$

Time Shifting:  $x[n - n_o] \Leftrightarrow z^{-n_o} X(z), n_o \geq 0$

$$x[n + n_o] \Leftrightarrow z^{n_o} \left( X(z) - \sum_{m=0}^{n_o-1} x[m]z^{-m} \right), n_o \geq 0$$